

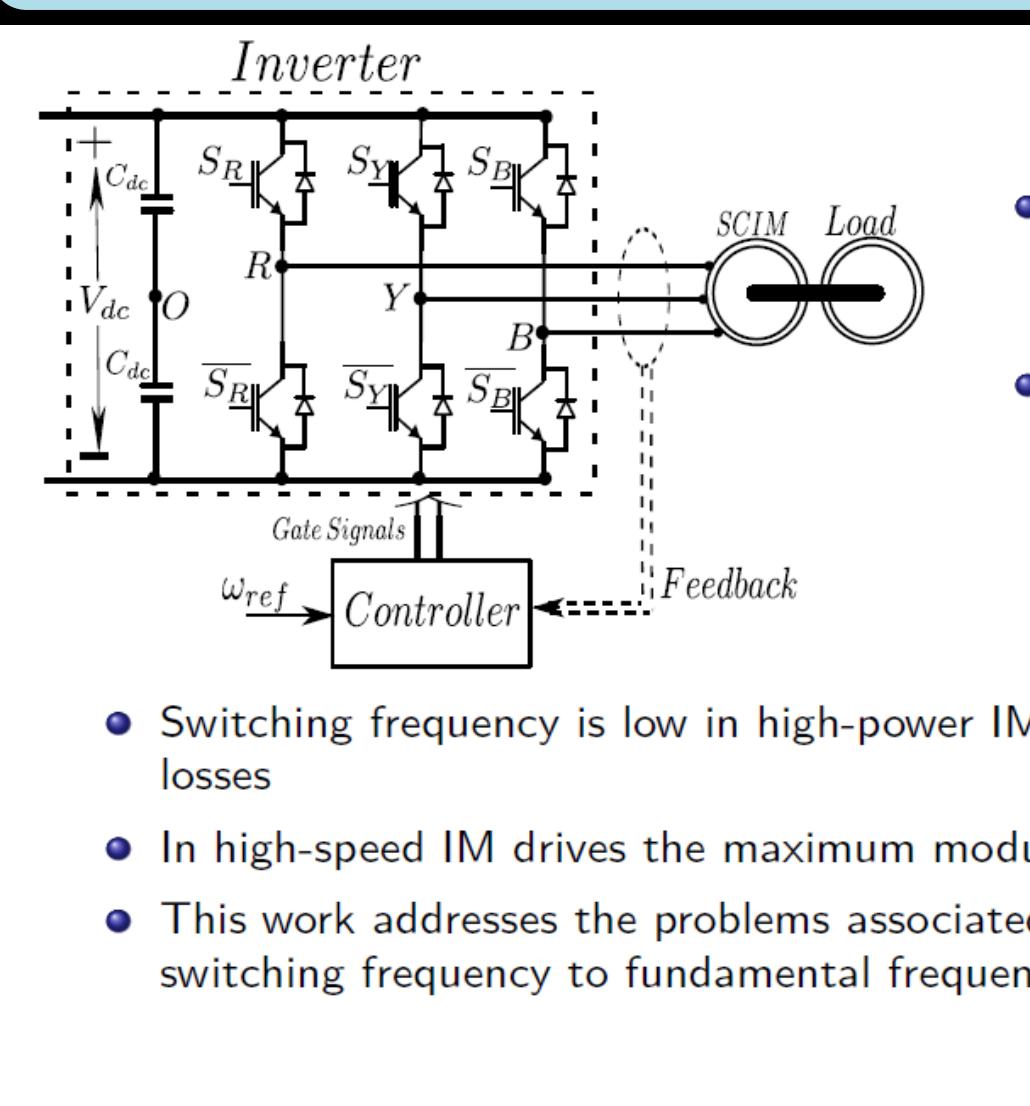
# LOW SWITCHING FREQUENCY PULSE WIDTH MODULATION FOR INDUCTION MOTOR DRIVES



AVANISH TRIPATHI, PROF. G NARAYANAN

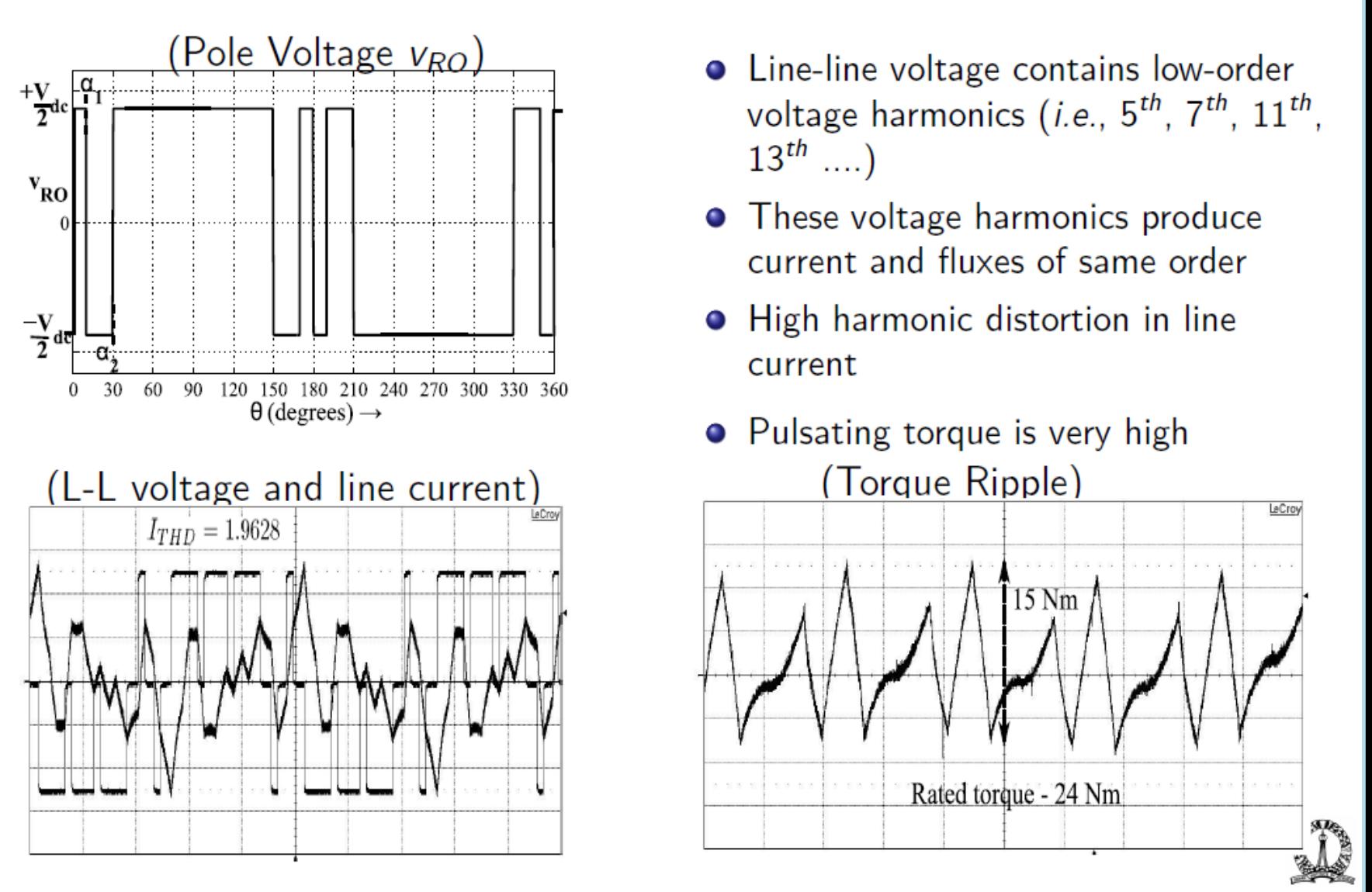
Department of Electrical Engineering Indian Institute of Science, Bangalore - 560012 INDIA  
avanish@ee.iisc.ernet.in, gnar@ee.iisc.ernet.in

## Introduction



- Variable-amplitude variable-frequency voltage is generated to control the induction motor
- Switching frequency is generally much higher than maximum modulation frequency
- Switching frequency is low in high-power IM drives due to high switching energy losses
- In high-speed IM drives the maximum modulation frequency is quite high
- This work addresses the problems associated with such cases where the ratio of switching frequency to fundamental frequency (pulse number,  $P$ ) is low

## Problems with Low Pulse Number Applications



- Line-line voltage contains low-order voltage harmonics (i.e., 5<sup>th</sup>, 7<sup>th</sup>, 11<sup>th</sup>, 13<sup>th</sup> ...)
- These voltage harmonics produce current and fluxes of same order
- High harmonic distortion in line current
- Pulsating torque is very high (Torque Ripple)

## Space Vector Analysis

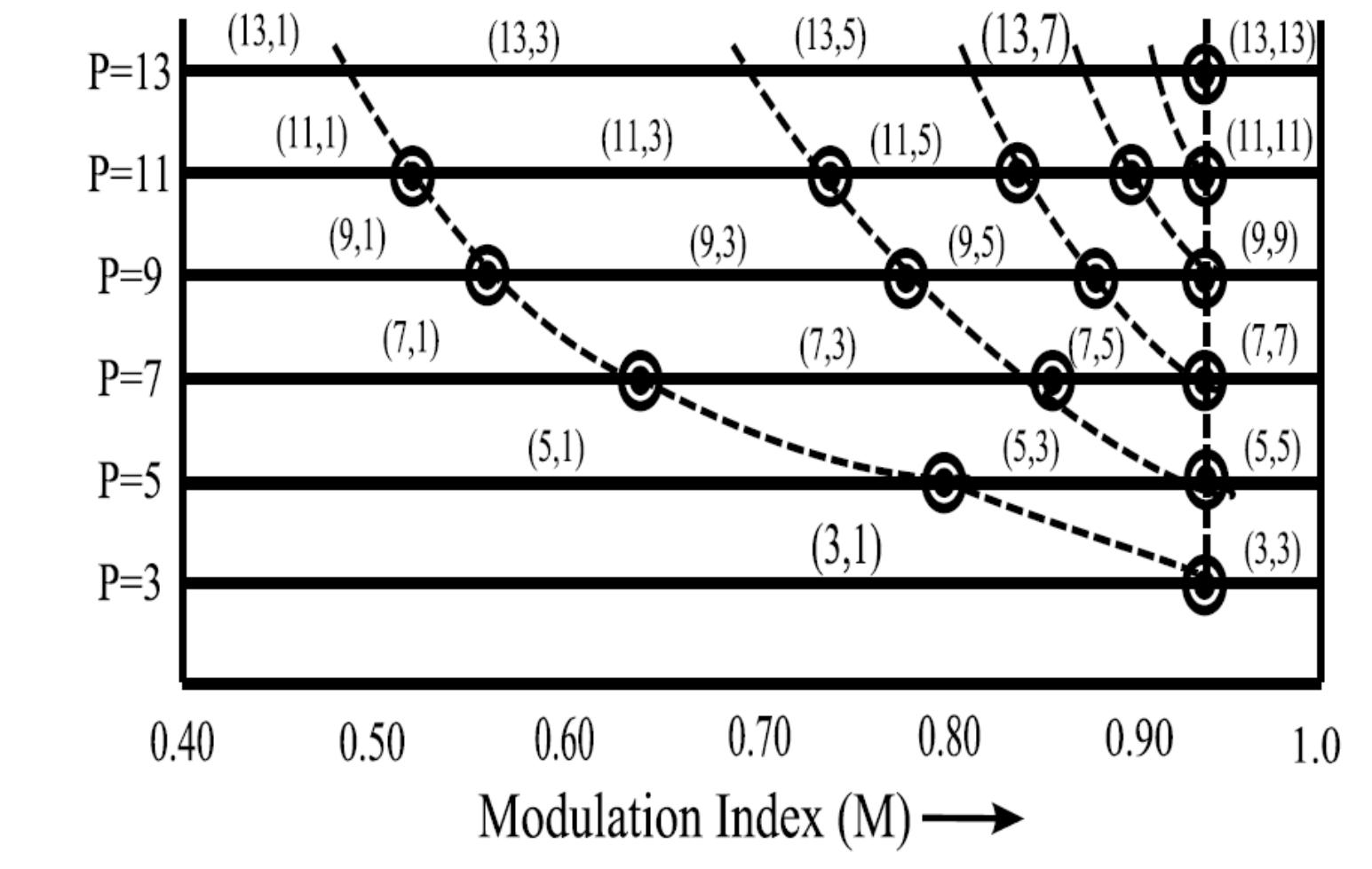
Modulation index (M)	Vector sequence in		Optimal sequence code
	Optimal Type A	Optimal Type B	
0 < M ≤ 0.8	101 - 272	012 - 127	101-272
0.8 < M ≤ 0.88	101 - 272	012 - 127	012 - 127
0.88 < M ≤ 0.94	121 - 212	012 - 127	012 - 127
0.94 < M < 1.0	121 - 212	012 - 127	121 - 212

Modulation index (M)	Optimal vector sequence	No. of transitions	Optimal sequence code	
			$N_{aa}$	$N_{az}$
0 < M ≤ 0.8	101-272	1	4	(5,1)
0.8 < M ≤ 0.94	012 - 127	3	2	(5,3)
0.94 < M < 1.0	121 - 212	5	0	(5,5)

- Only nearest active vectors are applied and initial state in sector I is either 0 or 1
- Zero states 7 and 0 are not applied in 1<sup>st</sup> and 2<sup>nd</sup> halves of sector I, respectively
- Number of active vector transitions ( $N_{aa}$ ) gradually increases with M
- Optimal sequence code  $\Rightarrow (P, N_{aa})$

## Optimal Switching Sequence Map

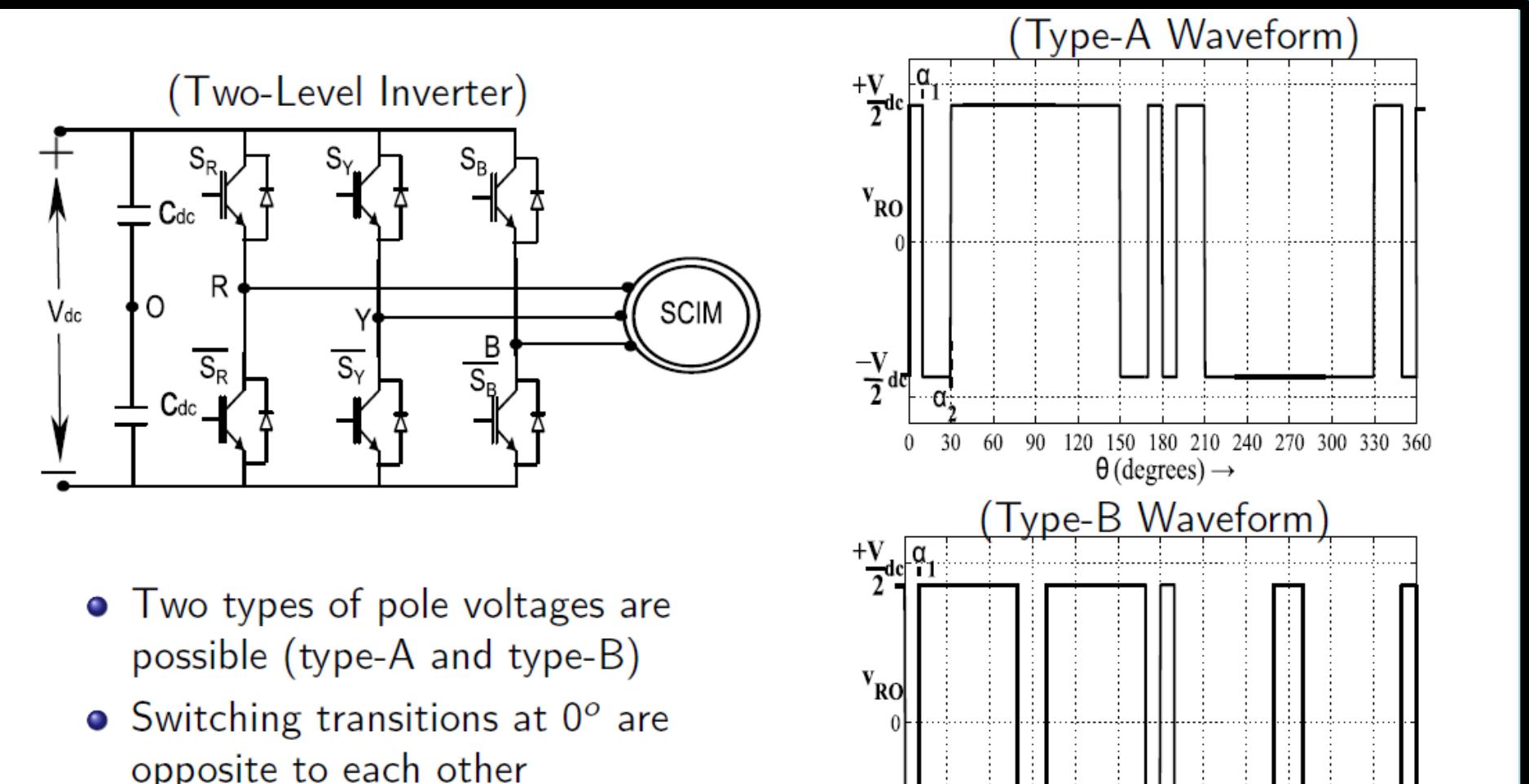
- Optimal sequence codes and the M ranges are plotted on a P Vs M map
- The curves can be extended to any pulse number



## Objectives

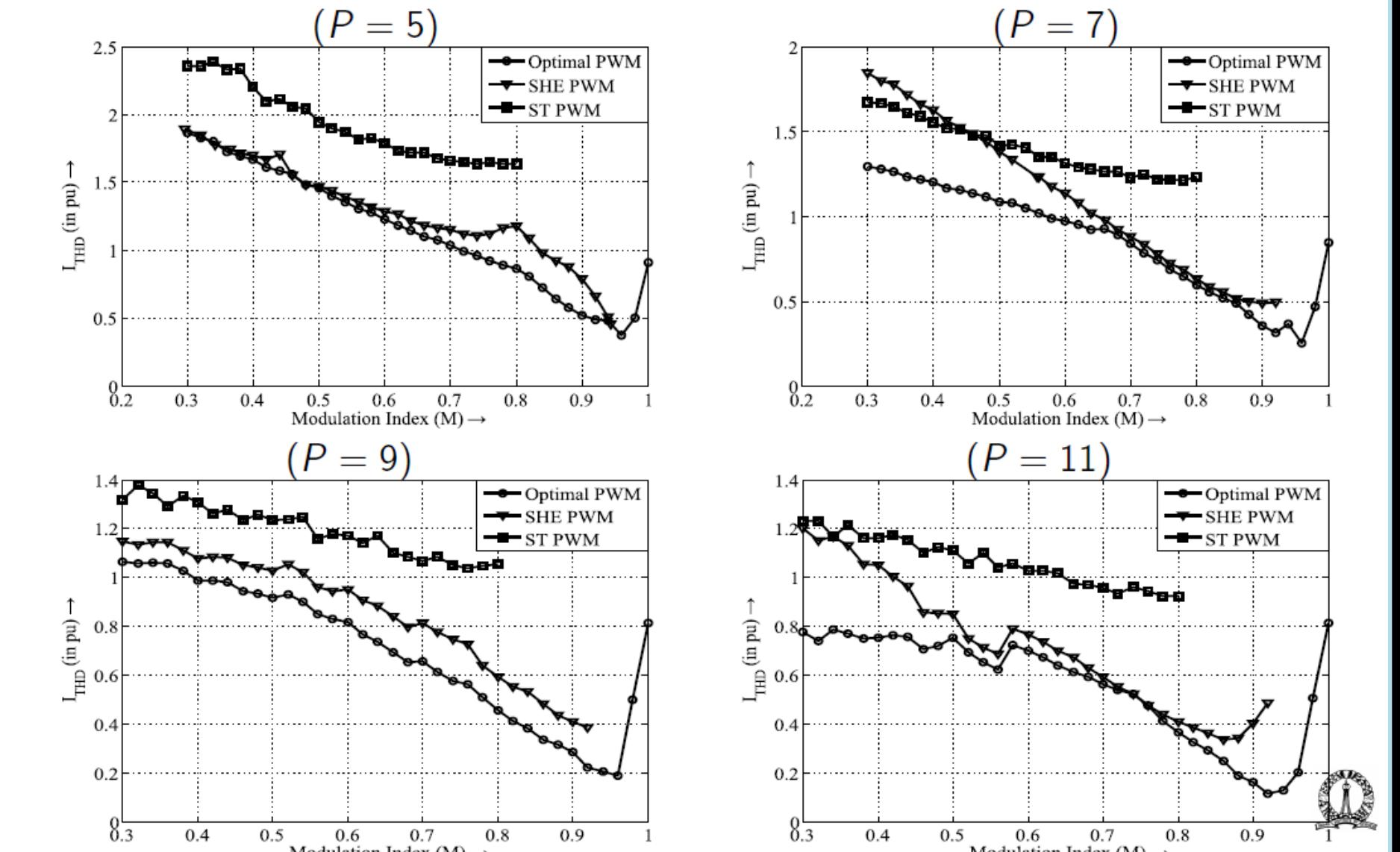
- Minimization of line current THD
- Determination of optimal switching sequences in space vector
- Minimization of pulsating torque in frequency domain and synchronous reference frame
- Extend the proposed scheme to neutral point clamped three-level inverter
- Predict the current and torque ripple based on PWM voltage
- Closed-loop control of IM drive operated with low switching frequency varying between 250Hz and 500Hz

## Current THD Minimization

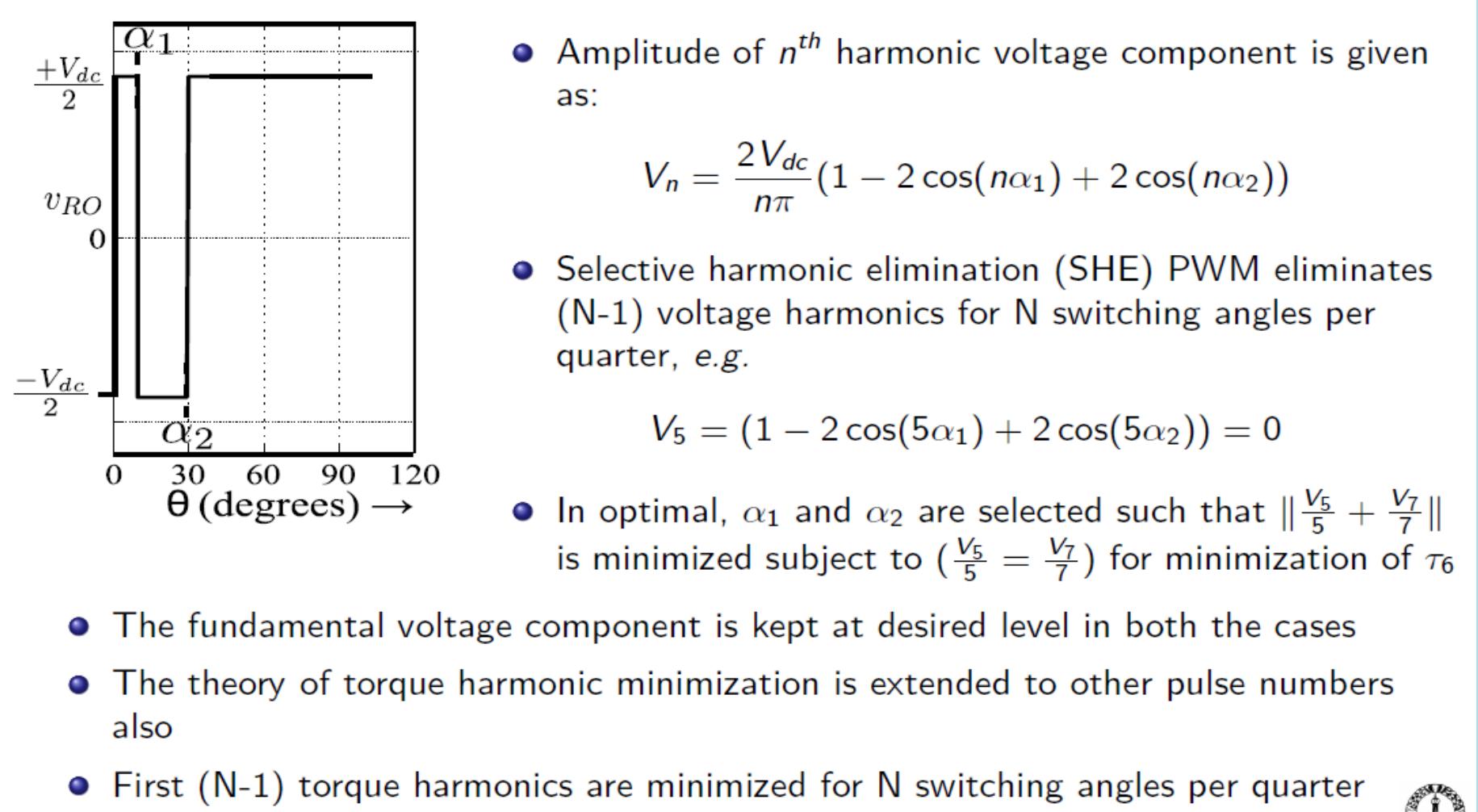


- Two types of pole voltages are possible (type-A and type-B)
- Switching transitions at 0° are opposite to each other
- Optimal PWM is solved for pulse numbers (P) 5, 7, 9 and 11

## Performance Evaluation in Terms of Current THD

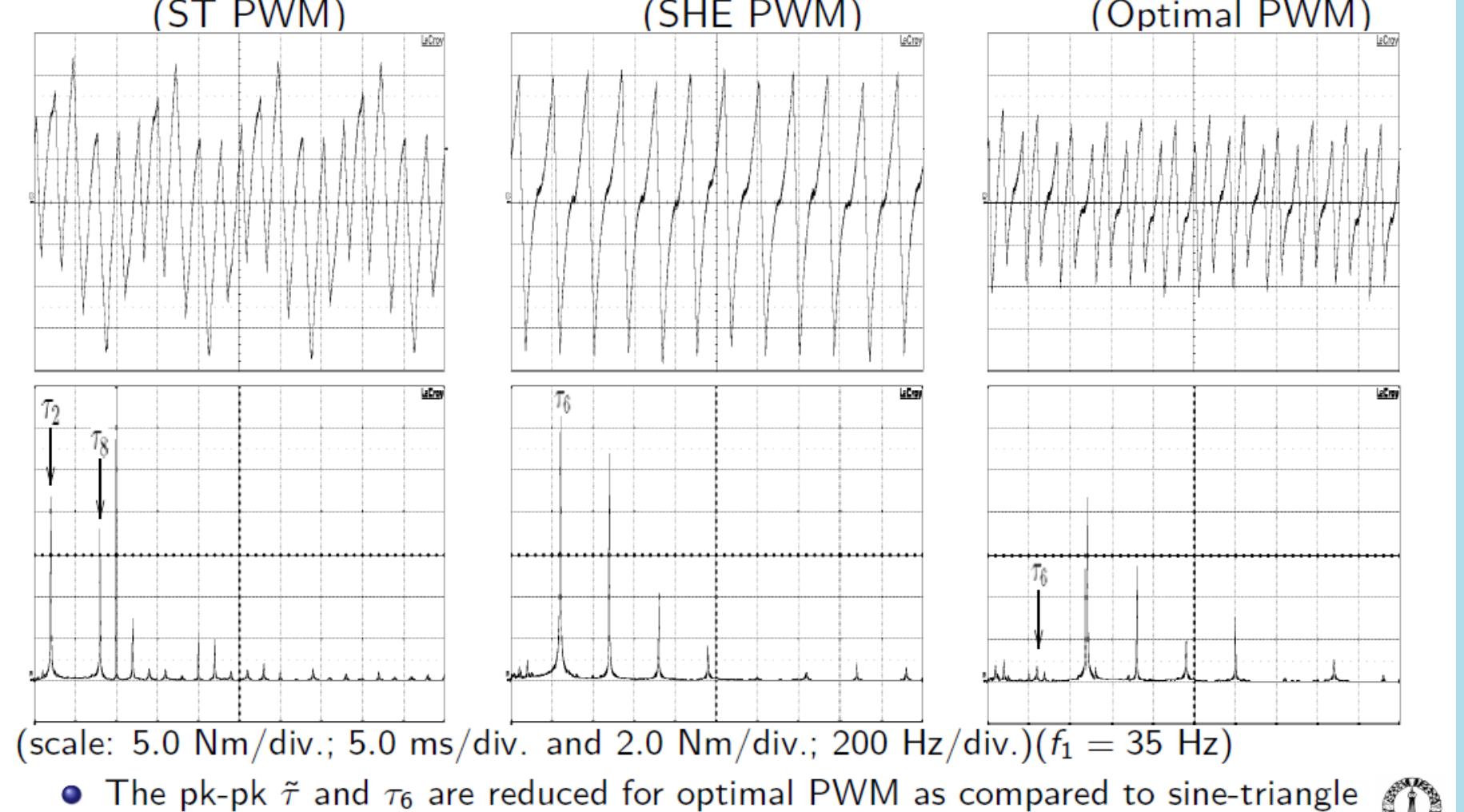


## Torque Ripple Minimization

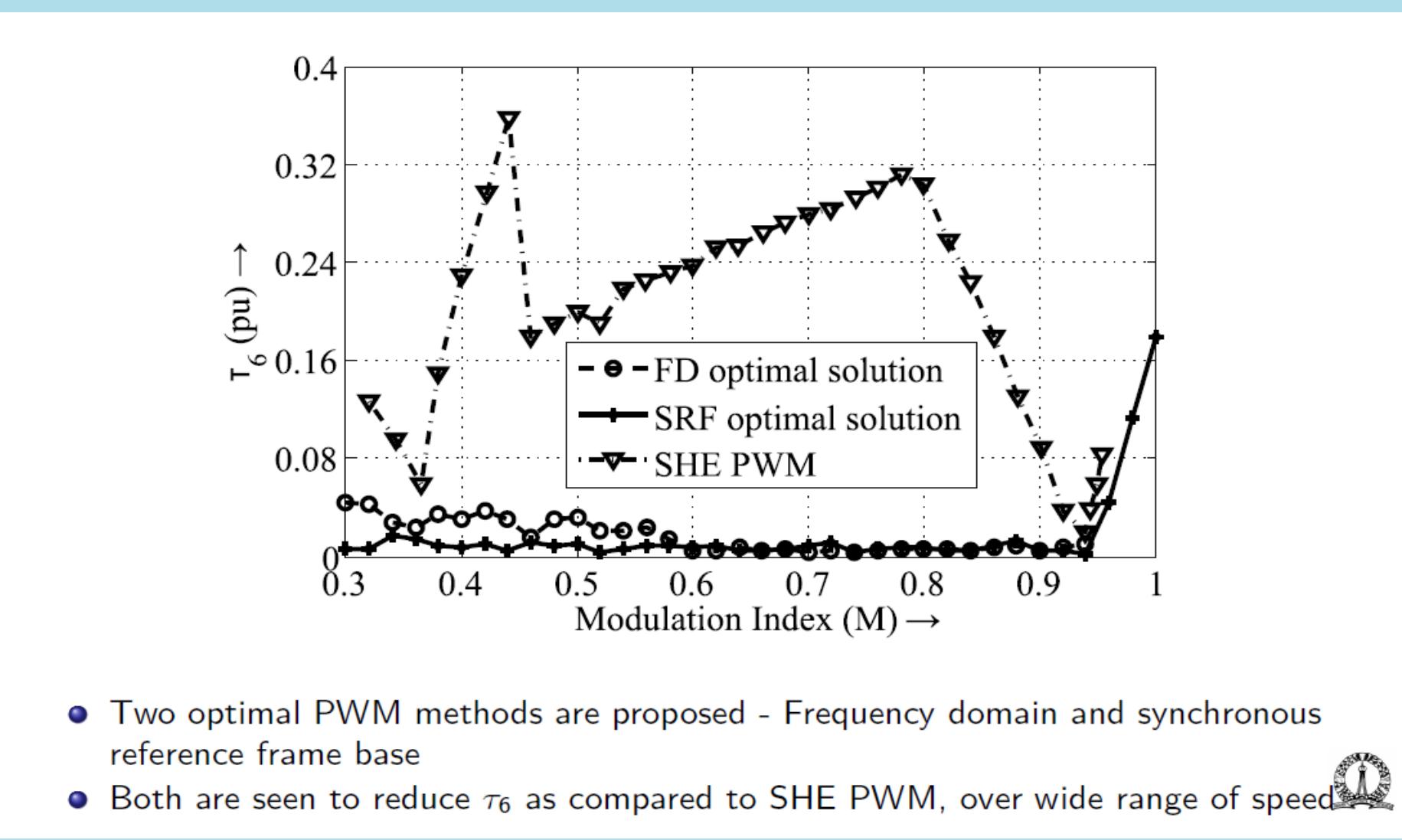


- Amplitude of  $n^{\text{th}}$  harmonic voltage component is given as:  $V_n = \frac{2V_{dc}}{n\pi} (1 - 2\cos(n\alpha_1) + 2\cos(n\alpha_2))$
- Selective harmonic elimination (SHE) PWM eliminates (N-1) voltage harmonics for N switching angles per quarter, e.g.  $V_5 = (1 - 2\cos(5\alpha_1) + 2\cos(5\alpha_2)) = 0$
- In optimal,  $\alpha_1$  and  $\alpha_2$  are selected such that  $\| \frac{V_5}{5} + \frac{V_7}{7} \|$  is minimized subject to  $(\frac{V_5}{5} = \frac{V_7}{7})$  for minimization of  $\tau_6$
- The fundamental voltage component is kept at desired level in both the cases
- The theory of torque harmonic minimization is extended to other pulse numbers also
- First (N-1) torque harmonics are minimized for N switching angles per quarter

## Comparison of Pulsating Torque For P=5



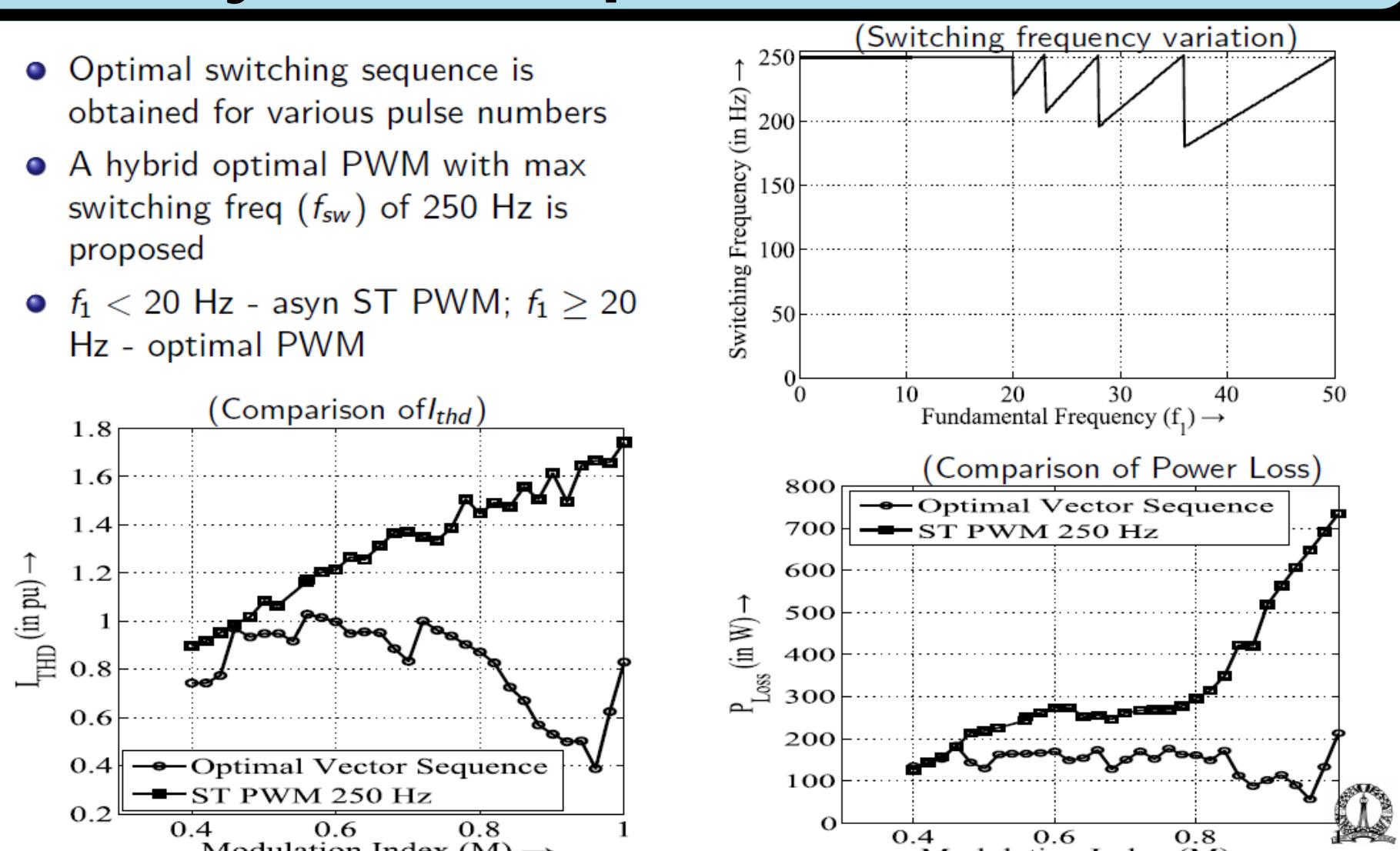
## Experimental Comparison of Harmonic Torque for P=5



- Two optimal PWM methods are proposed - Frequency domain and synchronous reference frame base
- Both are seen to reduce  $\tau_6$  as compared to SHE PWM, over wide range of speed

## Hybrid Optimal PWM

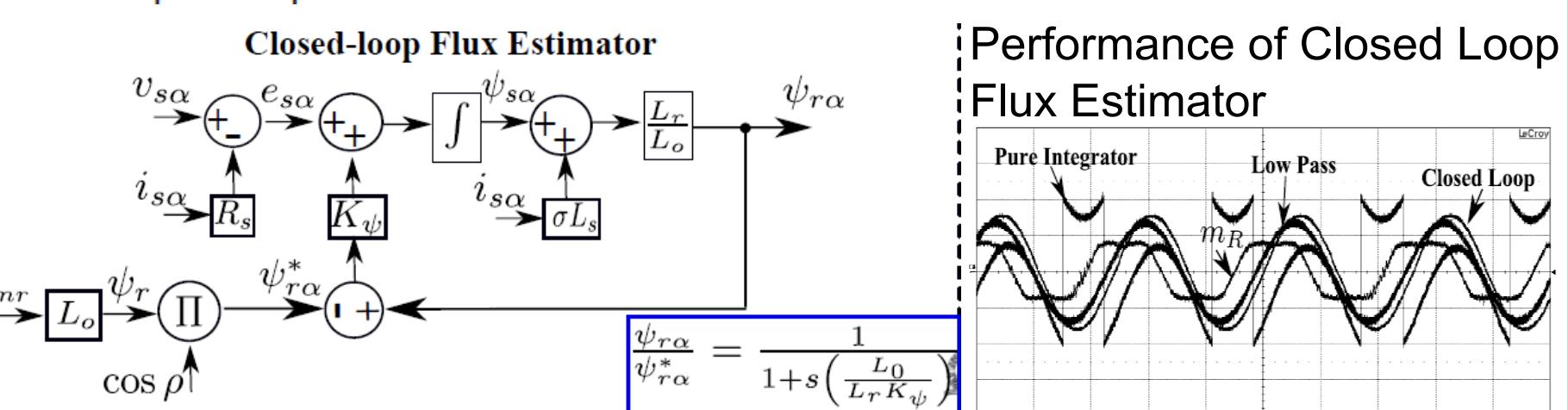
- Optimal switching sequence is obtained for various pulse numbers
- A hybrid optimal PWM with max switching freq ( $f_{sw}$ ) of 250 Hz is proposed
- $f_1 < 20$  Hz - asyn ST PWM;  $f_1 \geq 20$  Hz - optimal PWM



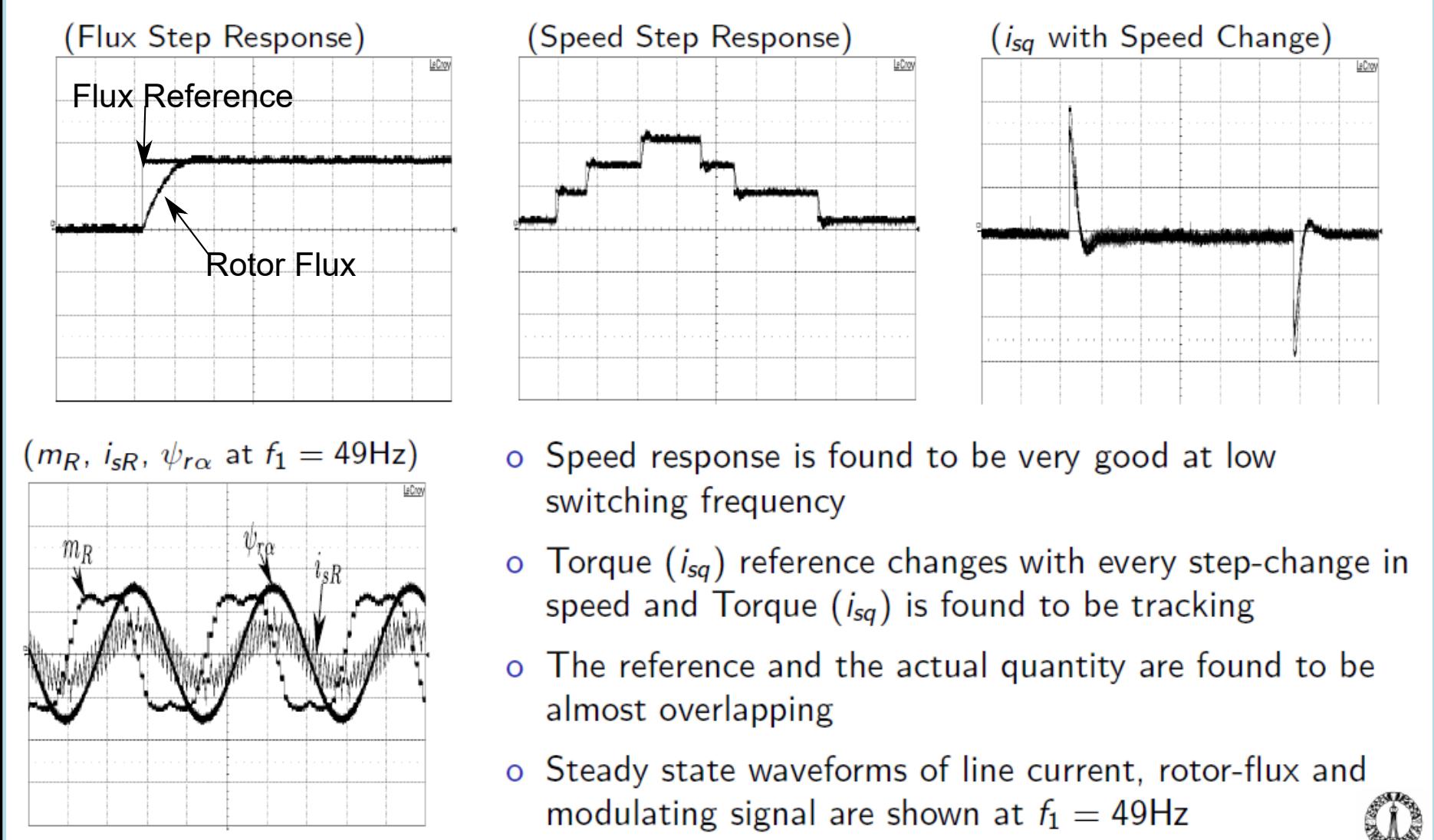
## Vector Control of IM

### Challenges with Low switching Frequency PWM

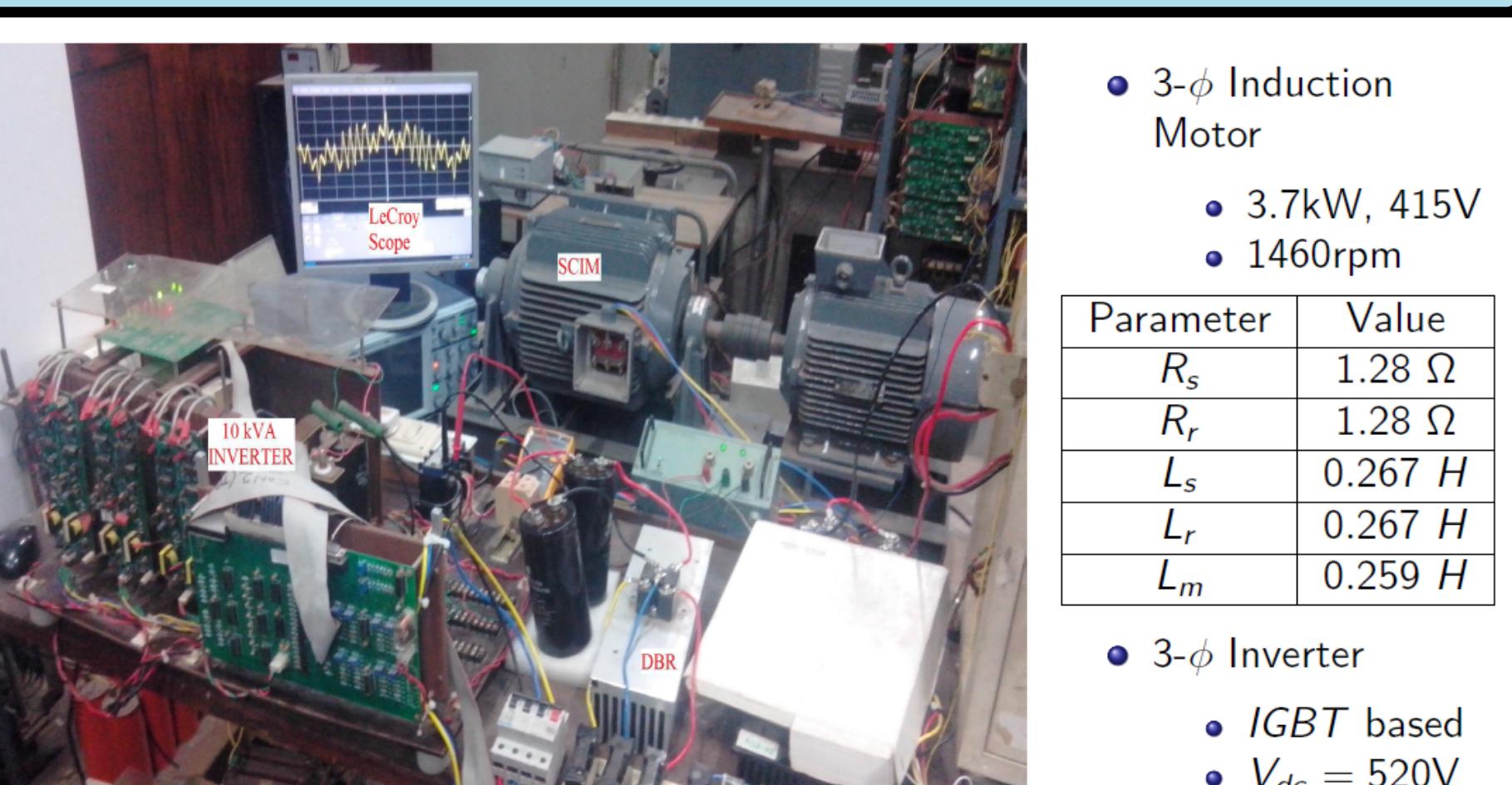
- Decoupled control of flux and torque is required for fast dynamics
- Too high current ripple  $\Rightarrow$  may destabilize the controller
- The inverter delay becomes significant
- The rotor flux estimation is required for sensorless control
- Flux estimation becomes inaccurate (phase and amplitude) due to low sampling frequency
- Closed loop flux estimator design is proposed and it is compared with open loop flux estimator



### Controller Performance Evaluation



## Experimental Setup



FPGA based controller board is used for open-loop V/f control

## Important Publications

- A. Tripathi and G. Narayanan, "Evaluation and minimization of low-order harmonic torque in low-switching-frequency inverter fed induction motor drives," *IEEE Trans. Ind. Appl.*, Mar 2015, Early access.
- A. Tripathi and G. Narayanan, "Evaluation and minimization of low-order harmonic torque in low-switching-frequency inverter fed induction motor drives," *IEEE PEDES*, Dec 2014, pp. 1-6.
- A. Tripathi and G. Narayanan, "Investigations on optimal pulse-width modulation to minimize total harmonic distortion in the line current," *IEEE IICPE*, Dec 2014, pp. 1-6.
- A. Tripathi and G. Narayanan, "High-Performance Off-line Pulse Width Modulation Without Quarter Wave Symmetry For Voltage-Source Inverter," *IEEE ICAECC*, Oct 2014, pp. 1-6.