### An Alternating Projections Algorithm for Sparse Blind Deconvolution

#### Aniruddha Adiga

Department of Electrical Engineering Indian Institute of Science, Bangalore Email: aniruddha@ee.iisc.ernet.in

Advisor: Prof. Chandra Sekhar Seelamantula





イロン イヨン イヨン イヨン

Problem Formulation: Blind Deconvolution (BD)

• Consider the linear measurement model

$$\mathbf{y} = \mathbf{h} \ast \mathbf{e} + \mathbf{w} = \mathbf{H}\mathbf{e} + \mathbf{w} = \mathbf{E}\mathbf{h} + \mathbf{w},$$

$$\begin{split} \mathbf{h} &\in \mathbb{R}^{L}: \text{ Point spread function (PSF).} \\ \mathbf{e} &\in \mathbb{R}^{M}: \text{ Excitation.} \\ \mathbf{y} &\in \mathbb{R}^{N}: \text{ observation.} \\ \mathbf{w} &\in \mathbb{R}^{N}: \text{ Noise } (N = L + M - 1). \\ \mathbf{E} &\in \mathbb{R}^{N \times L}: \text{ linear convolution matrices corresponding to } \mathbf{e}. \\ \mathbf{H} &\in \mathbb{R}^{N \times M}: \text{ linear convolution matrices corresponding to } \mathbf{h}. \end{split}$$

イロト 不得 トイヨト イヨト

# Some Examples

• Debluring: Blurring filter not known.



2

・ロト ・回ト ・ヨト ・ヨト

### Some Examples

• Debluring: Blurring filter not known.



• Point source blurred due Gaussian PSF with unknown parameters.



<ロ> (日) (日) (日) (日) (日)

### Some Examples

• Debluring: Blurring filter not known.



• Point source blurred due Gaussian PSF with unknown parameters.



• Linear speech production model: Factorize voiced speech into vocal-tract filter and excitation.



# • Develop a sparse blind deconvolution (SBD) algorithm to estimate h and e from y.

Assumption: **e** is a sparse vector and **h** is a smooth and stable operator.

э

# • Develop a sparse blind deconvolution (SBD) algorithm to estimate h and e from y.

Assumption: **e** is a sparse vector and **h** is a smooth and stable operator.

э

- Objective
   Develop a sparse blind deconvolution (SBD) algorithm to estimate h and e from y.
   Assumption: e is a sparse vector and h is a smooth and stable operator.
- Typical issues Infinitely many solutions to the BD problem. Need priors on e and/or h to reduce search space.

3

# **Objective** • Develop a sparse blind deconvolution (SBD) algorithm to estimate **h** and **e** from **y**.

Assumption: **e** is a sparse vector and **h** is a smooth and stable operator.

#### **Typical issues**

- issues Infinitely many solutions to the BD problem. Need priors on e and/or h to reduce search space.
  - The formulated cost function is non-convex, thus leading to local minima issues.

#### • We formulate the SBD problem using the MAP formulation.

- Propose an alternating minimization algorithm for the problem.
- We show that for well-conditioned systems, the pseudo-inverse solution is a good initialization.
- We analyze the convergence property of the algorithm.
- We show application to deconvolution of voiced speech signals.

- We formulate the SBD problem using the MAP formulation.
- Propose an alternating minimization algorithm for the problem.
- We show that for well-conditioned systems, the pseudo-inverse solution is a good initialization.
- We analyze the convergence property of the algorithm.
- We show application to deconvolution of voiced speech signals.

(日) (周) (王) (王)

- We formulate the SBD problem using the MAP formulation.
- Propose an alternating minimization algorithm for the problem.
- We show that for well-conditioned systems, the pseudo-inverse solution is a good initialization.
- We analyze the convergence property of the algorithm.
- We show application to deconvolution of voiced speech signals.

(日) (周) (王) (王)

- We formulate the SBD problem using the MAP formulation.
- Propose an alternating minimization algorithm for the problem.
- We show that for well-conditioned systems, the pseudo-inverse solution is a good initialization.
- We analyze the convergence property of the algorithm.
- We show application to deconvolution of voiced speech signals.

(日) (周) (王) (王)

- We formulate the SBD problem using the MAP formulation.
- Propose an alternating minimization algorithm for the problem.
- We show that for well-conditioned systems, the pseudo-inverse solution is a good initialization.
- We analyze the convergence property of the algorithm.
- We show application to deconvolution of voiced speech signals.

・ロン ・四 と ・ ヨ と ・

#### MAP Formulation for BD problem

- h: deterministic but unknown.
- e: *i.i.d.* generalized p-Gaussian distribution (gpG).

$$f(\mathbf{e}) = \left(\frac{p}{2\Gamma(1/p)\gamma\sigma_e}\right)^M \exp\left(-\sum_i \left(\frac{|e_i|}{\gamma\sigma_e}\right)^p\right)$$

where  $\gamma = \left(\frac{\Gamma(1/p)}{\Gamma(3/p)}\right)^{1/2}$  and  $0 \le p \le 1$  (Heavy-tailed distribution).

• The MAP estimates of the vectors **h** and **e** are

$$\begin{aligned} (\mathbf{h}_{\mathsf{MAP}}, \mathbf{e}_{\mathsf{MAP}}) &= \arg \max_{\mathbf{h}, \mathbf{e}} \ f(\mathbf{y}/\mathbf{e}; \mathbf{h}) f(\mathbf{e}), \\ &= \arg \min_{\mathbf{h}, \mathbf{e}} \ \underbrace{\|\mathbf{y} - \mathbf{He}\|_2^2 + \lambda \|\mathbf{e}\|_p^p}_{F(\mathbf{h}, \mathbf{e})}. \end{aligned}$$

- The joint cost *F*(**h**, **e**) is non-convex and not straightforward to optimize.
- We resort to alternating minimization.



<ロ> (日) (日) (日) (日) (日)

э

An Alternating  $\ell_p - \ell_2$  Projections Algorithm (ALPA) I



- 4 同 6 4 日 6 4 日 6

An Alternating  $\ell_p - \ell_2$  Projections Algorithm (ALPA) II

Iteratively reweighted least-squares (IRLS)

whe

• Majorize  $\|\mathbf{e}\|_p^p$  with weighted  $\ell_2$ -norm function and minimize the cost iteratively.

$$\begin{split} \tilde{\mathbf{e}}^{(j+1,k)} &= \arg\min_{\mathbf{e}} \ \underbrace{\|\mathbf{y} - \mathbf{H}^{(k)}\mathbf{e}\|_{2}^{2} + \lambda \mathbf{e}^{\mathsf{T}} \mathbf{W}^{j,k} \mathbf{e}}_{F_{\epsilon}(\mathbf{h}^{(k)},\mathbf{e})} \end{split}$$
  
re  $\mathbf{W}^{(j,k)} &= \operatorname{diag}\left( p\left( (\tilde{\mathbf{e}}_{i}^{(j,k)})^{2} + \epsilon \right)^{p/2-1} \right). \end{split}$ 

• Final e-step solution (after J iterations of IRLS):  $e^{(k+1)} = \tilde{e}^{(J,k)}$ ,

$$\mathbf{e}^{(k+1)} = \left( (\mathbf{H}^{(k)})^T (\mathbf{H}^{(k)}) + \lambda \mathbf{W}^{(J,k)} \right)^{-1} (\mathbf{H}^{(k)})^T \mathbf{y},$$

• Use  $\mathbf{e}^{(k+1)}$  to update  $\mathbf{h}^{(k+1)} = \arg\min_{\mathbf{h}} F_{\epsilon} \left( \mathbf{h}, \mathbf{e}^{(k+1)} \right)$ . **h-step solution**:  $\mathbf{\tilde{h}}^{(k+1)} = \arg\min_{\mathbf{h}} \|\mathbf{y} - \mathbf{E}^{(k+1)}\mathbf{h}\|_{2}^{2} = \mathbf{E}^{(k+1)^{\dagger}}\mathbf{y}$ ,

$$\mathbf{h}^{(k+1)} = \tilde{\mathbf{h}}^{(k+1)} / \|\tilde{\mathbf{h}}^{(k+1)}\|_2^2.$$

• A good initialization might ensure we are in the "right" local minima.

э

- A good initialization might ensure we are in the "right" local minima.
- Initialization: Given some initial filter estimate  $\tilde{\mathbf{h}}$ , how good is the regularized least-squares solution (  $\mathbf{W}^{(0,0)} = \mathbf{1}$ )?

イロト 不得 トイヨト イヨト

- A good initialization might ensure we are in the "right" local minima.
- Initialization: Given some initial filter estimate  $\tilde{h}$ , how good is the regularized least-squares solution (  $W^{(0,0)}=1)?$
- If true excitation and filter are  $e^*$  and  $h^*$  respectively,  $\delta h = h^* \tilde{h}$  and pseudo-inverse solution is denoted  $\hat{e}_{\text{BLS}}$ . The difference

$$\delta \mathbf{e}_{\mathsf{BLS}} = \mathbf{e}^* - \hat{\mathbf{e}}_{\mathsf{BLS}} = \mathbf{H}^{*\dagger} (\mathbf{w} - \delta \mathbf{H} \mathbf{e}^*),$$

where  $\delta \mathbf{H} = \text{Conv. mtx}(\delta \mathbf{h})$ .

< ロ > < 同 > < 回 > < 回 > < □ > <

- A good initialization might ensure we are in the "right" local minima.
- Initialization: Given some initial filter estimate  $\tilde{h}$ , how good is the regularized least-squares solution (  $W^{(0,0)}=1)?$
- If true excitation and filter are  $e^*$  and  $h^*$  respectively,  $\delta h = h^* \tilde{h}$  and pseudo-inverse solution is denoted  $\hat{e}_{\text{BLS}}$ . The difference

$$\delta \mathbf{e}_{\mathsf{BLS}} = \mathbf{e}^* - \hat{\mathbf{e}}_{\mathsf{BLS}} = \mathbf{H}^{*\dagger} (\mathbf{w} - \delta \mathbf{H} \mathbf{e}^*),$$

where  $\delta \mathbf{H} = \text{Conv. mtx}(\delta \mathbf{h})$ .

• What is the probability of  $\frac{1}{M} \|\delta \mathbf{e}_{BLS}\|_1$  exceeding a value  $\xi$ ?

イロト 不得下 イヨト イヨト 二日

- A good initialization might ensure we are in the "right" local minima.
- Initialization: Given some initial filter estimate  $\tilde{h}$ , how good is the regularized least-squares solution (  $W^{(0,0)}=1)?$
- If true excitation and filter are  $e^*$  and  $h^*$  respectively,  $\delta h = h^* \tilde{h}$  and pseudo-inverse solution is denoted  $\hat{e}_{\text{BLS}}$ . The difference

$$\delta \mathbf{e}_{\mathsf{BLS}} = \mathbf{e}^* - \hat{\mathbf{e}}_{\mathsf{BLS}} = \mathbf{H}^{*\dagger} (\mathbf{w} - \delta \mathbf{H} \mathbf{e}^*),$$

where  $\delta \mathbf{H} = \text{Conv.} \text{mtx}(\delta \mathbf{h})$ .

• What is the probability of  $\frac{1}{M} \|\delta \mathbf{e}_{\mathsf{BLS}}\|_1$  exceeding a value  $\xi$ ?

$$P\left(\frac{1}{M}\|\delta \mathbf{e}_{\mathsf{BLS}}\|_1 > \xi\right) \leq \frac{\kappa^2 \sigma^2}{(\xi - \kappa \|\delta \mathbf{H}\|_2 \|\mathbf{e}^*\|_2 / \sqrt{M})^2}$$

 $\kappa =$ condition number of  $\mathbf{H}^*$ .



Behaviour of the majorized cost function  $F_{\epsilon}(\mathbf{h}, \mathbf{e})$ :

2

・ロト ・回ト ・ヨト ・ヨト

Behaviour of the majorized cost function  $F_{\epsilon}(\mathbf{h}, \mathbf{e})$ :

• After e-step,

$$F_{\epsilon}(\mathbf{h}^{(k)}, \mathbf{e}^{(k+1)}) \leq F_{\epsilon}\left(\mathbf{h}^{(k)}, \mathbf{e}^{(k)}\right).$$

2

Behaviour of the majorized cost function  $F_{\epsilon}(\mathbf{h}, \mathbf{e})$ :

• After e-step,

$$F_{\epsilon}(\mathbf{h}^{(k)}, \mathbf{e}^{(k+1)}) \leq F_{\epsilon}\left(\mathbf{h}^{(k)}, \mathbf{e}^{(k)}\right).$$

• Similarly, after h-step,

$$\mathcal{F}_{\epsilon}(\mathbf{h}^{(k+1)},\mathbf{e}^{(k)}) \leq \mathcal{F}_{\epsilon}\left(\mathbf{h}^{(k)},\mathbf{e}^{(k)}
ight).$$

э

・ロト ・回ト ・ヨト ・ヨト

Behaviour of the majorized cost function  $F_{\epsilon}(\mathbf{h}, \mathbf{e})$ :

• After e-step,

$$F_{\epsilon}(\mathbf{h}^{(k)}, \mathbf{e}^{(k+1)}) \leq F_{\epsilon}\left(\mathbf{h}^{(k)}, \mathbf{e}^{(k)}
ight).$$

• Similarly, after h-step,

$$F_{\epsilon}(\mathbf{h}^{(k+1)}, \mathbf{e}^{(k)}) \leq F_{\epsilon}(\mathbf{h}^{(k)}, \mathbf{e}^{(k)}).$$

• After one cycle of ALPA, (both e-step and h-step)

$$F_{\epsilon}(\mathbf{h}^{(k+1)}, \mathbf{e}^{(k+1)}) \leq F_{\epsilon}(\mathbf{h}^{(k)}, \mathbf{e}^{(k)}).$$

Behaviour of the majorized cost function  $F_{\epsilon}(\mathbf{h}, \mathbf{e})$ :

• After e-step,

$$\mathcal{F}_{\epsilon}(\mathbf{h}^{(k)},\mathbf{e}^{(k+1)}) \leq \mathcal{F}_{\epsilon}\left(\mathbf{h}^{(k)},\mathbf{e}^{(k)}
ight).$$

• Similarly, after h-step,

$$F_{\epsilon}(\mathbf{h}^{(k+1)}, \mathbf{e}^{(k)}) \leq F_{\epsilon}\left(\mathbf{h}^{(k)}, \mathbf{e}^{(k)}\right).$$

• After one cycle of ALPA, (both e-step and h-step)

$$F_{\epsilon}(\mathbf{h}^{(k+1)},\mathbf{e}^{(k+1)}) \leq F_{\epsilon}(\mathbf{h}^{(k)},\mathbf{e}^{(k)}).$$

Behaviour of the original Cost function  $F(\mathbf{h}, \mathbf{e})$ :

< ロ > < 同 > < 回 > < 回 > < 回 > <

Behaviour of the majorized cost function  $F_{\epsilon}(\mathbf{h}, \mathbf{e})$ :

• After e-step,

$$\mathcal{F}_{\epsilon}(\mathbf{h}^{(k)},\mathbf{e}^{(k+1)}) \leq \mathcal{F}_{\epsilon}\left(\mathbf{h}^{(k)},\mathbf{e}^{(k)}
ight).$$

• Similarly, after h-step,

$$F_{\epsilon}(\mathbf{h}^{(k+1)}, \mathbf{e}^{(k)}) \leq F_{\epsilon}\left(\mathbf{h}^{(k)}, \mathbf{e}^{(k)}\right).$$

• After one cycle of ALPA, (both e-step and h-step)

$$F_{\epsilon}(\mathbf{h}^{(k+1)}, \mathbf{e}^{(k+1)}) \leq F_{\epsilon}(\mathbf{h}^{(k)}, \mathbf{e}^{(k)}).$$

Behaviour of the original Cost function  $F(\mathbf{h}, \mathbf{e})$ :

• With  $\epsilon$  updated as  $\epsilon^{(k)} = c \left( \max \left| \mathbf{e}_j^{(k)} \right| \right)^{2-p}$ , with  $0 < c \ll 1$ ,

$$F\left(\mathbf{h}^{(k+1)}, \mathbf{e}^{(k+1)}\right) \leq F\left(\mathbf{h}^{(k)}, \mathbf{e}^{(k)}\right).$$

イロト 不得 トイヨト イヨト

#### Deconvolution of Voiced Speech Signal



• 30 ms vowel segment /ae/ (female speaker),

Figure: (a) Speech samples (sampling rate 16 kHz); (b) LP residue (MO 20); (c) ALPA estimate of the filter; (d) ALPA estimate of the excitation, (e) frequency response of the estimated filter; and (f) comparison between the original vowel segment shown in (a) and that synthesized based on the estimated filter and excitation.

Comparisons with Smooth  $\ell_1/\ell_2$  blind deconvolution (SOOT)<sup>1</sup>, sparse linear prediction (SLP)<sup>2</sup> and MM-based Sparse Deconvolution (SDMM)<sup>3</sup>.

<sup>&</sup>lt;sup>1</sup>A. Repetti et al. "Euclid in a Taxicab: Sparse Blind Deconvolution with Smoothed  $\ell_1/\ell_2$  Regularization". In: *IEEE Signal Process.* Lett. 22.5 (2015), pp. 539–543. DOI: 10.1109/LSP.2014.2362861. eprint: 1407.5465.

<sup>&</sup>lt;sup>2</sup>D. Giacobello et al. "Sparse Linear Prediction and Its Applications to Speech Processing". In: IEEE Trans. Audio, Speech, Language Process. 20.5 (2012), pp. 1644–1657. ISSN: 1558-7916. DOI: 10.1109/TASL.2012.2186807.

<sup>&</sup>lt;sup>3</sup>Ivan Selesnick. Sparse Deconvolution (An MM Algorithm).

http://cnx.org/contents/f2738de6-b36d-458d-a2dd-2b50f375fe5505/Sparse-Deconvolution-An=MM-Alg 🗇 🕨 🗧 🐑 🛓 🔗 🔍

#### Comparisons with state-of-the-art



Figure: A comparison of sparse deconvolution methods for clean speech: Rows 1, 2, and 3 corresponding to Column 1 show the speech signal, frequency response of the LP filter, and the LP residue, respectively. For Columns 2–4, Rows 1, 2, and 3 show estimates of the filter, its frequency response, and the excitation, respectively.

э

<ロ> (日) (日) (日) (日) (日)

# Summary



• We considered the problem of sparse blind deconvolution.

2

・ロン ・四 と ・ ヨ と ・ ヨ と

- We considered the problem of sparse blind deconvolution.
- MAP formulation of the problem using super-Gaussian priors for a sparse solution.

3

・ロト ・回ト ・ヨト ・ヨト

# Summary

- We considered the problem of sparse blind deconvolution.
- MAP formulation of the problem using super-Gaussian priors for a sparse solution.
- Resulting cost function is non-convex and was optimized using an alternating  $\ell_p \ell_2$  projections algorithm (ALPA).
- ALPA breaks the optimization problem into two convex sub problems for estimation of excitation and filter.

3

< ロ > < 同 > < 回 > < 回 > < □ > <

# Questions?

2

・ロン ・四 と ・ ヨ と ・ ヨ と