



HIGH VOLTAGE POWER SUPPLY AND CROWBAR PROTECTION

Subhash Joshi T.G., CDAC Trivandrum
Vinod John, IISc Bangalore

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Objective of research and applications

Design analysis and experimental validation of

- i. Solid State Crowbar (SSC) for the high voltage high power applications
- ii. High performance High Voltage Power Supply (HVPS) for medium power applications

➤ **Applications of SSC**

- ✓ Impulse current generator
- ✓ Electromagnetic launcher
 - ❖ strong magnetic field produced by discharging capacitor
- ✓ Partial discharge detection
 - ❖ damped AC voltage is produced
- ✓ Electro-magnetic forming
 - ❖ produce electromagnetic forces to achieve shaping, joining, cutting
- ✓ **Protection for Microwave tube (MWT) used in plasma generation**

➤ **Applications of HVPS**

- ✓ Industry
 - ❖ Electrostatic precipitator for filtering fine particles
 - ❖ Corona generators
 - ❖ Dielectric testing
- ✓ Medical
 - ❖ **X-rays**
- ✓ Strategic
 - ❖ **Radar**
 - ❖ Particle accelerators
 - ❖ Plasma applications

Need for HV pulse power electronics

- System performance in the applications depends on output voltage ripple
 - Image contrast in X-ray and MWT are affected
- Survivability of Microwave Tube (MWT) found in HV applications depends on
 - Stored energy in the HVPS
- HVPS is classified based on the power rating

High power

Topology: *Mains frequency rectifier circuit*

- ✓ Large output capacitors used to reduce the ripple
- ✓ Increase the stored energy
 - ✓ demands protection mechanism called **Crowbar**

Medium power

Topology: *Switched converters*

- ✓ Reduce the effect of low frequency input ripple on the output voltage called **Audio susceptibility**

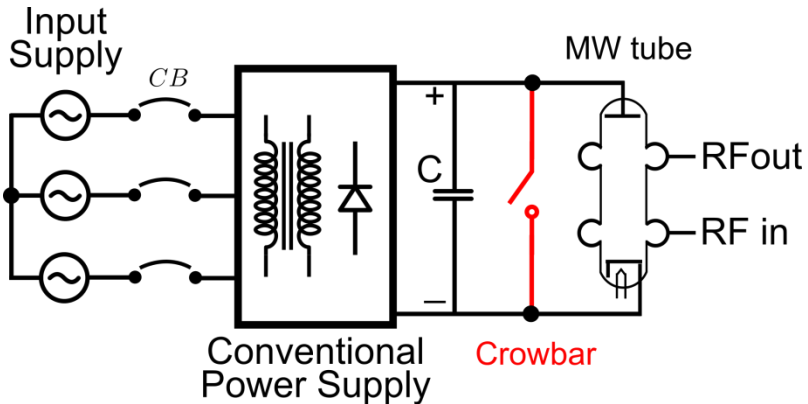
Low power

Topology: *Voltage multiplier circuits*

Solid state crowbar (SSC)

➤ Requirement of Solid State Crowbar (SSC)

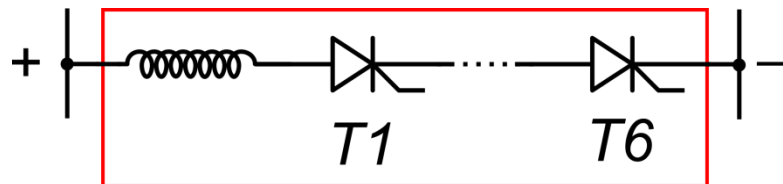
- Limit the energy in MW tube below 10J



➤ Commonly used switches are

- Ignitron, thyatron [1-3]
 - mercury gas filled tube
 - use of these devices is restricted now a days
- spark gap [4]
 - nitrogen gas filled device
 - frequent inspection and maintenance

➤ State of art is to use semiconductor devices [5]



- SSC of 10kV dc, 1kA peak current and 10MW peak power
- Should transfer energy in 10 μ s



Contributions of the research work on SSC

1. Modelling of
 - ✓ DC fault current in MW tube
 - ✓ Modelling of MW tube
2. Design of di/dt limiting inductor
3. Static and dynamic voltage balancing network
4. Design of mounting clamps for press pack thyristor
5. Other topics
 - Computation of transient thermal impedance for the thyristor and heatsink combined system
 - Selection of cable for the pulse power application

1. Contributions for modelling

➤ Model of fault current:

- To match the **Joules Integral (J.I.)** due to the current instead matching the actual waveform

$$J.I. = \int i_d^2(t) dt$$

- Fault current is due to the current from input supply and due to the discharge of output capacitor,

$$i_d(t) = I_{d,avg} \left[1 - e^{-\delta t} \left(\cos \omega_d t + \frac{\delta}{\omega_d} \sin \omega_d t \right) \right] + \frac{V_{C,init}}{R_1} e^{-\frac{t}{\tau_1}}$$

➤ Model of MW tube:

- Internal arc event is modeled using a fuse wire
- Energy conservation relation for the fuse wire for an incremental time Δt

$$i_d^2(t) R(\sigma(T)) \Delta t = A l \rho C_P \Delta T + d\psi_{conduction} + d\psi_{convection} + d\psi_{radiation}$$

- J.I. and temperature are related as

$$\int_0^t i_d^2(\tau) d\tau = \frac{A^2 \rho C_P \sigma_o}{\alpha_o} \ln(1 + \alpha_o (T - T_o))$$

2. Design of di/dt limiting inductor

- Required to avoid local hot spot in the thyristor due to high di/dt
- Energy accumulation in the MW tube is increased by,
 - di/dt limiting inductor
 - Turn-on delay time of the thyristor
- Two boundary conditions of inductance are estimated from,
 - Expression relating inductance, energy in MW tube, and turn on delay time
 - Expression relating inductance and slope of current through thyristor (di/dt)



Inductance: 250 μ H

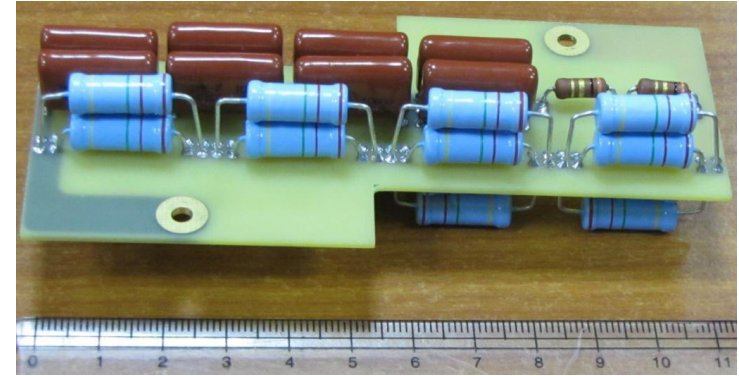
Voltage: 10kV dc

Peak current rating: 1kA

WxDia: 32x12cm

3. Voltage balancing network

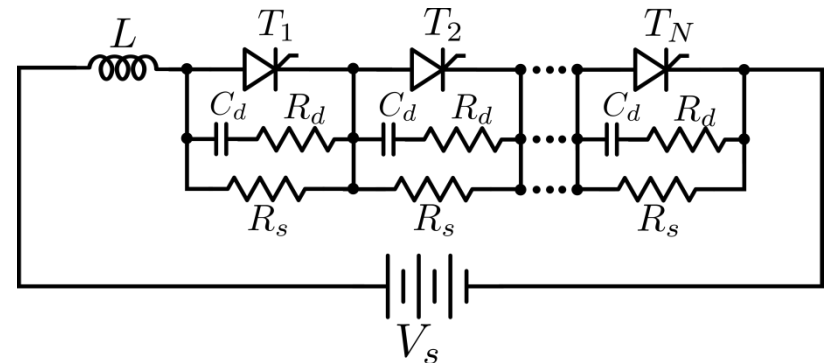
- Design based on difference in turn-on delay
- Component tolerances are considered
- Differences in propagation delays among thyristor drivers are incorporated
- Functionality without complex pulse synchronizing circuit
- Expressions for static and dynamic components



WxHxD: 11x3x5.5cm

$$V_{C_d, \max} = \frac{V_s}{N} + \frac{V_s(N-1)}{Nt_{on}} \left[t_{dON} - \sqrt{L(1-a_C)C_d} \sin\left(\frac{t_{dON}}{\sqrt{L(1-a_C)C_d}}\right) \right] ; \quad R_d C_d = t_{on} / 50$$

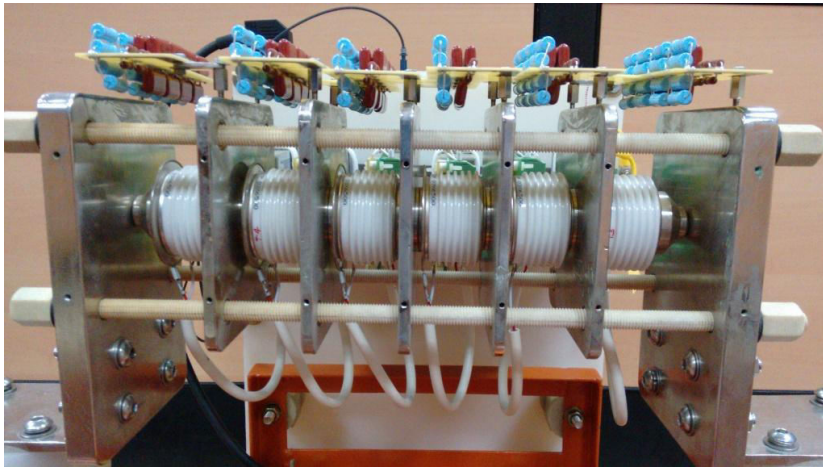
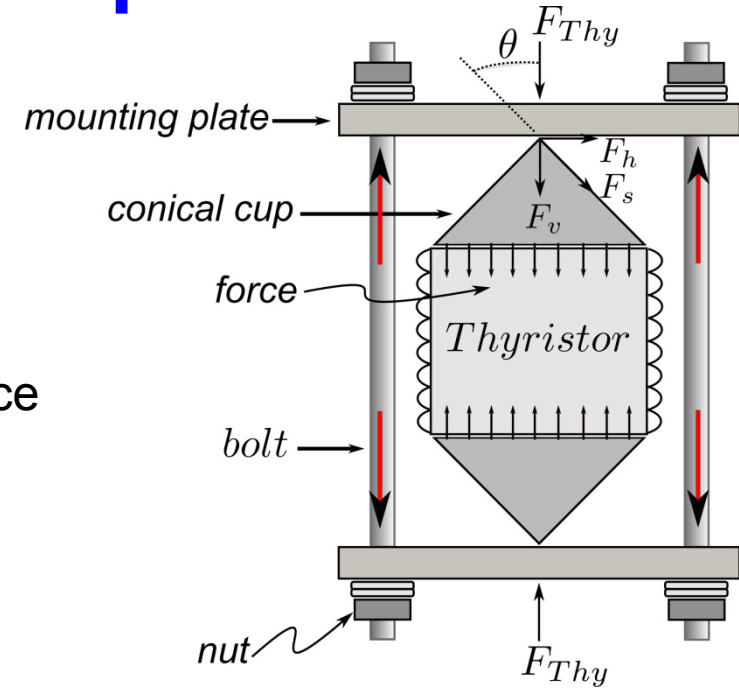
Variable	Method	
	Proposed	Conventional
C_d	40nF	2250nF (56 times)
I_{\max}	16A	900A (56 times)



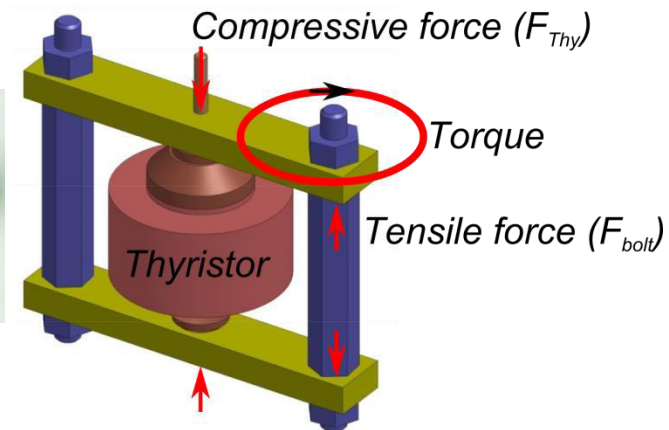
1. Indian patent filed with number "6679/CHE/2015 dated 12/12/2015" jointly by CDAC & IISc
 2. T. G. Subhash Joshi and V. John, "Static and Dynamic Balancing Network for Crowbar Application," in Proc. National Power Electronics Conference, Bombay, Dec. 2015.

4. Mounting clamp

- Mounting force required for the desired electrical characteristic from the thyristor datasheet
- Design of mounting clamp involves:
 - Estimation of torque for the required force
 - Uniform force distribution over the pole face
 - Selection of bolt-nut and, design of mounting plate



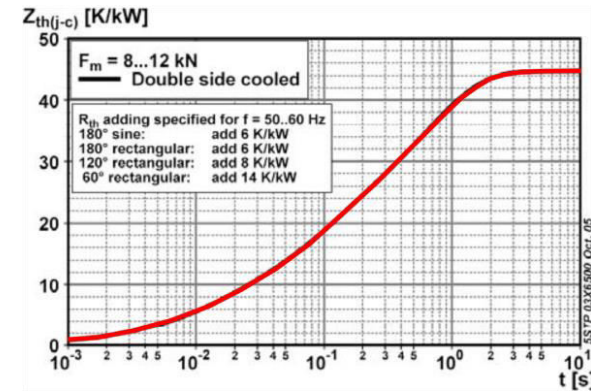
WxHxD: 33x22x12cm



5. Other contributions in SSC

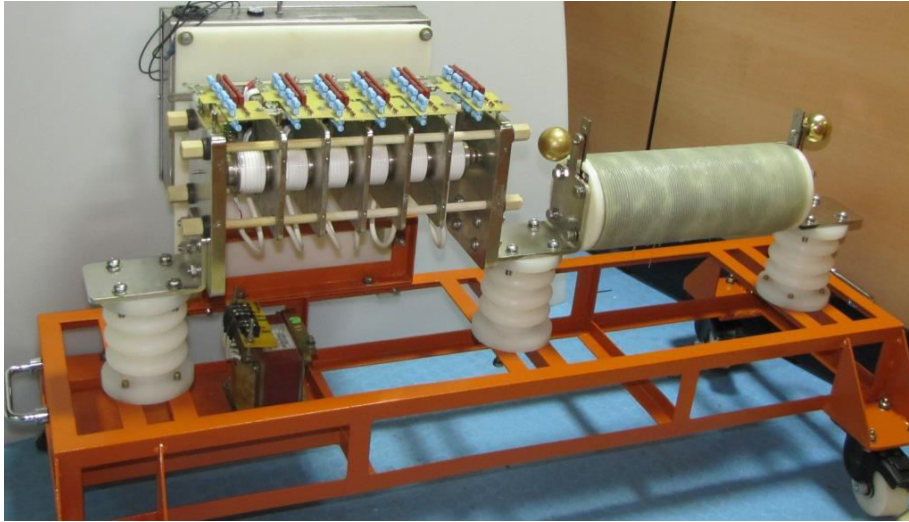
- **Computation of combined transient thermal impedance (Z_{th}) of the system**
 - Z_{th} captures the temperature variation at any point for a step input power
 - Z_{th} of subsystems are known
 - Thyristor , Heatsink: *can be estimated experimentally*
 - Z_{th} of the combined system is to be estimated

- **Selection of cable for the pulse power application**
 - High peak current flows for a short duration of time
 - Conventional method of selecting the cable by current density is inappropriate
 - Procedure considers:
 - ambient temperature
 - temperature rise in the cable insulation



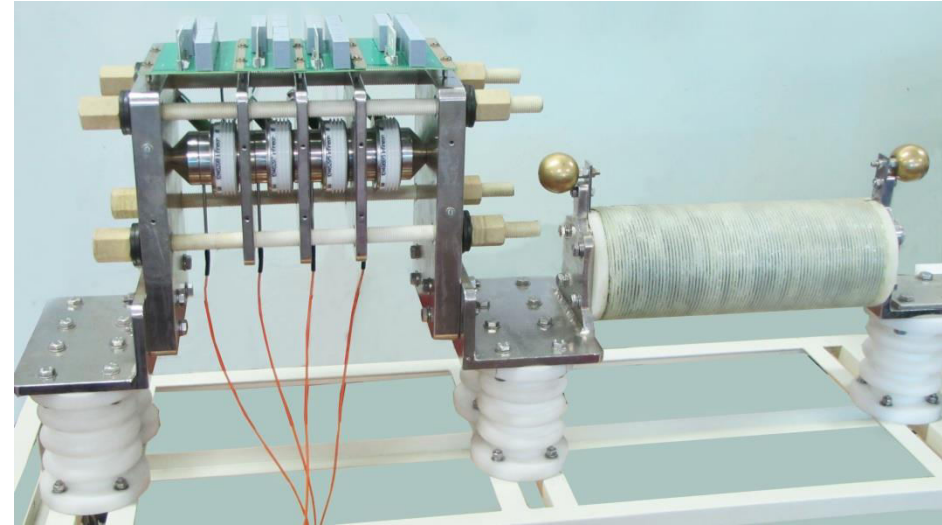
Solid State Crowbar

ETT based Crowbar



Dimension (WxHxD):
1.1m x 0.56m x 0.45m

LTT based Crowbar



Dimension (WxHxD):
1.0m x 0.65m x 0.53m

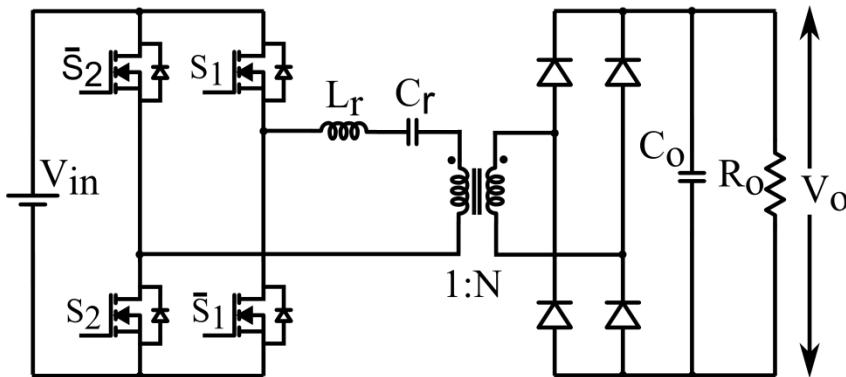
- *This research work led to the first MW power level Solid State Crowbar built in India*
 - Only one manufacturer worldwide for high power SSC (M/s JEMA, Spain)
- Supplied SSC with Light Triggered Thyristor to *Institute for Plasma Research (IPR), Ahmedabad*, where it is currently in operation

High Voltage Power Supply (HVPS)

➤ Requirement of High Voltage Power Supply

➤ $V_o = 10\text{kV}$, $P_o = 10\text{kW}$,

$E_s \leq 10\text{J}$, $\Delta V_o / V_o \leq 0.5\%$



➤ Storage energy and output ripple are related

- Switching at high frequency
- Series Resonant Converter (SRC) used
- $f_s = 108\text{kHz}$, $f_r = 100\text{kHz}$, $F = f_s / f_r$

➤ $Q = Z_c / R_o$, $Z_c = \sqrt{L_r / C_r}$

➤ Output ripple contribution from

- Switching action
- Input source ripple
 - Audio Susceptibility (AuS)

➤ Several modeling methods reported [6]

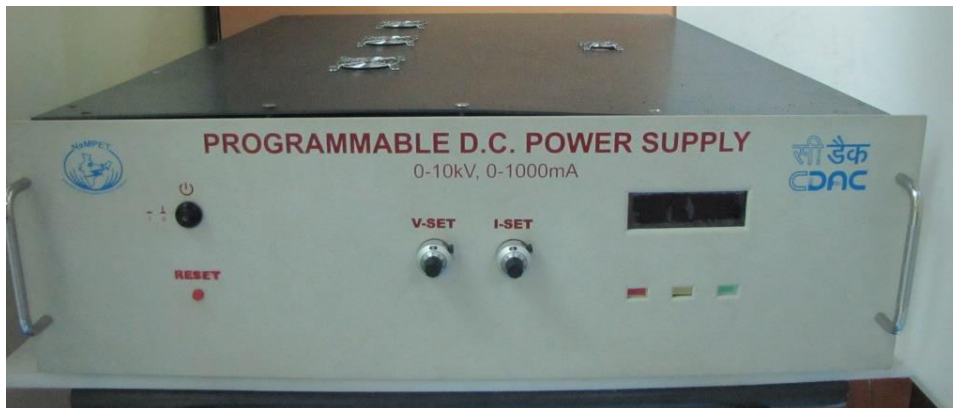
- fundamental approximation [7-8]
- numerically solved
 - hampers the design purpose

➤ AuS small signal models are not widely studied [9]

- Available models are numerically solved

Contributions of the research work on HVPS

1. AuS small signal model of SRC
 - Sampled data modelling method is used [10]
2. Design of SRC
 - minimize the effect of input ripple on output voltage
 - for superior audio susceptibility gain
3. High Voltage High Frequency (HVHF) magnetic design
 - for minimum stray capacitance
 - ensures resonant frequency above switching frequency



Dimension (WxHxD):
0.59mx0.18mx0.65m

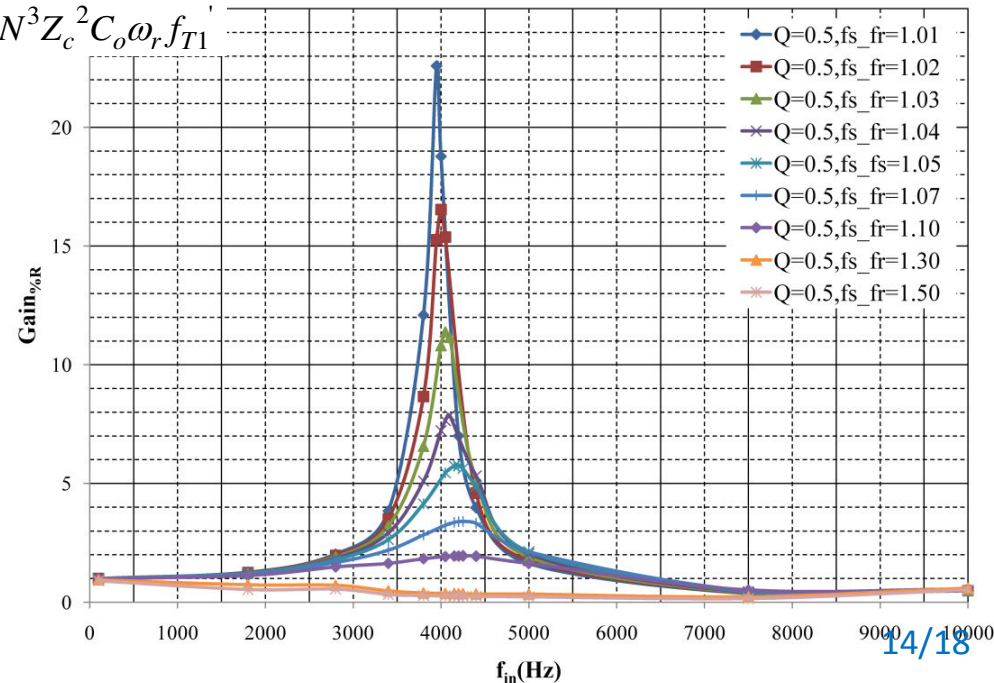
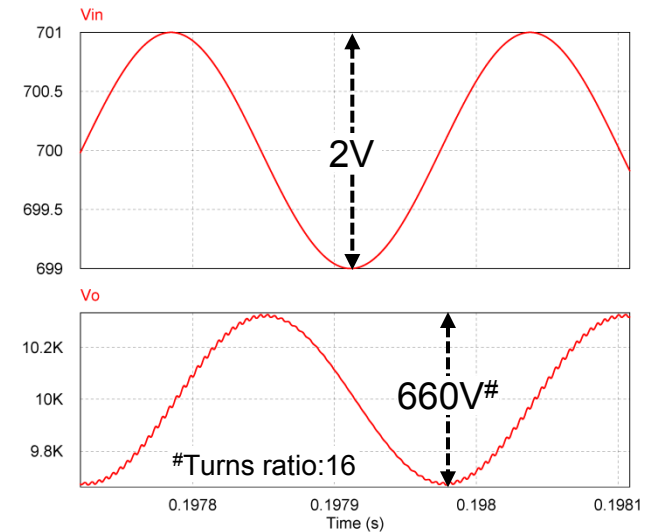
1. AuS small signal model

- Modeling involves estimation of:
 - Discrete time state space model using three state variables
 - Cyclic steady state of variables
 - Small signal AuS model and simplification based on practical conditions

$$\frac{v_o(z)}{v_{in}(z)} = \frac{\frac{16}{NZ_c C_o \omega_r} \left\{ (z-1) - \frac{4V_o}{NZ_c f_{T1}} \right\}}{(z-1)^3 - \frac{4V_o}{NZ_c f_{T1}} (z-1)^2 + \frac{16}{N^2 Z_c C_o \omega_r} (z-1) - \frac{64V_o}{N^3 Z_c^2 C_o \omega_r f_{T1}}}$$

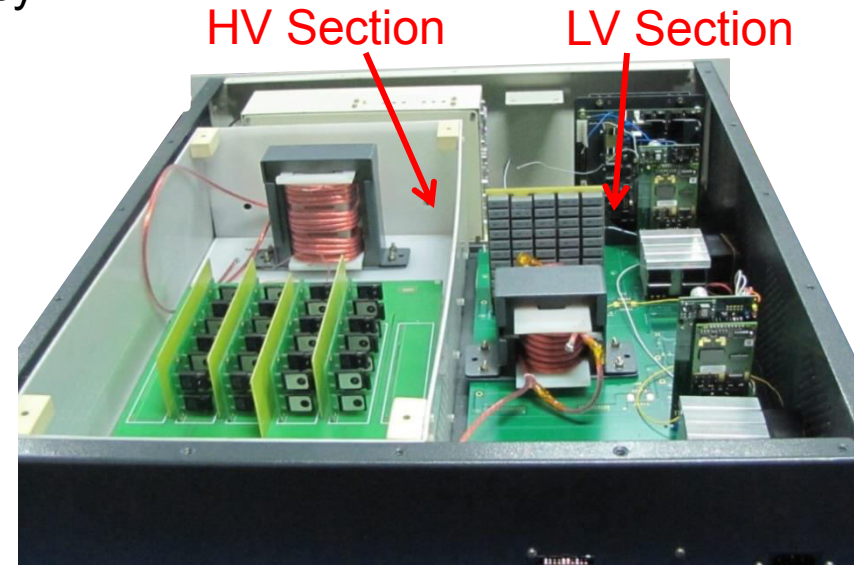
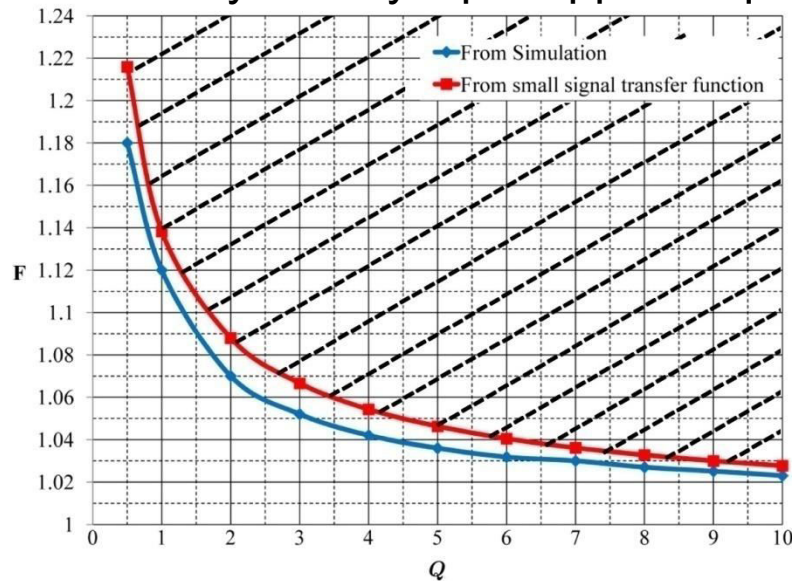
- Analysis shows
 - Resonance for the frequency response, gain as high as 22
 - Frequency where resonance occurs

$$\omega_{in,r} = \frac{1}{T_s} \tan^{-1} \left(\sqrt{\frac{16}{N^2 C_o \omega_r Z_c}} \right)$$



2. Design of SRC

- AuS analysis shows
 - there exists a set of values for F and Q where gain is always less than unity for any input ripple frequency

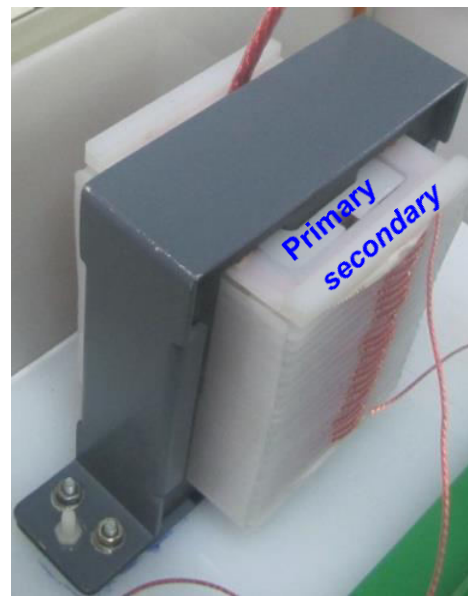
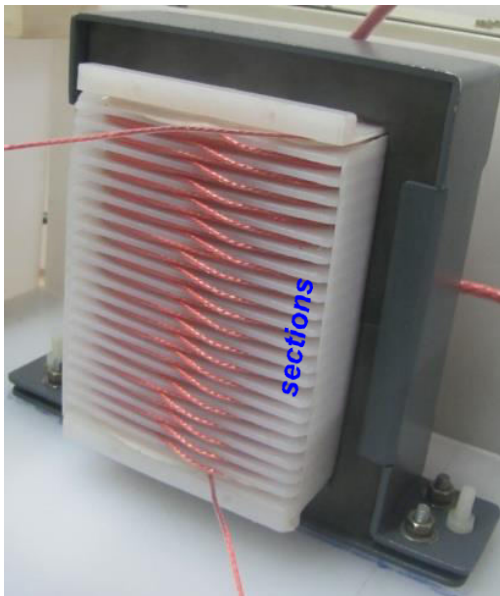
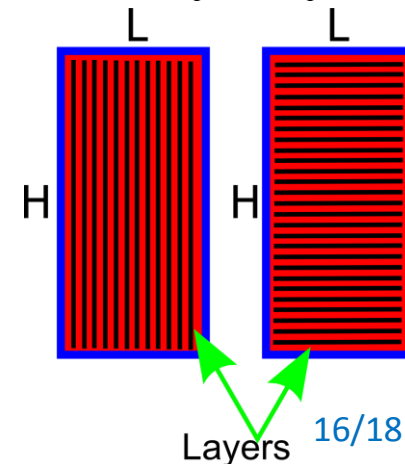
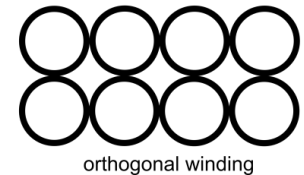
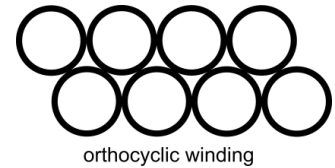
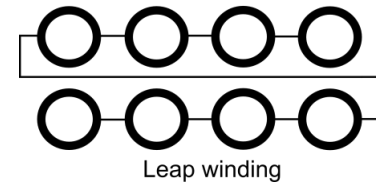
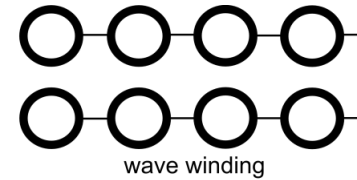


Parameter selection	Conventional method	Proposed method
Q and F	Q=0.65 to 4 F=1.05 to 1.2	Selected based on audio susceptibility gain is less than 1
Turns ratio and tank components	Fundamental approximation	Sampled data modelling

3. HVHF transformer

- Stray capacitance (C_s) is predominant at high frequency
- Experimental iterations

Sl. No.	Winding configuration	C_s	f_r
1	Wave, 0 section, 4layer	63.18pF	76.8kHz
2	Leap, 0 section, 4layer	52.95pF	82.8kHz
3	Leap, 3 section, 4layer	9.60pF	193.5kHz
4	Leap, 7 section, 8layer	8.00pF	216.5kHz
5	Leap, 21 section, 5layer	6.40pF	227.5kHz



Voltage: 700V/10kV
Power: 10kVA
Frequency: 160kHz
WxHxD: 15x12x8cm

Conclusions

- ❖ Crowbar for high power applications
 - Protection device analyzed and verified for energy sensitive load
 1. Modelling of DC fault current and MW tube
 2. Design of di/dt limiting Inductor
 3. Static and dynamic voltage balancing network
 4. Design of mounting clamps for crowbar application
 5. Method for computation of transient thermal impedance for crowbar and method for cable selection
- ❖ HVPS for medium power applications
 - Output voltage ripple minimized by proper selection of resonant components without increasing the stored energy
 1. Audio susceptibility small signal model of SRC
 2. Design of SRC
 3. High Voltage High Frequency magnetic design
- ❖ All the proposed methods are validated on hardware systems



THANK YOU

References

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9. S. Tian, F. Lee, and Q. Li, “A simplified equivalent circuit model of series resonant converter,” IEEE Trans. Power Electronics, vol. 31, no. 5, pp. 3922–3931, 2016.
10. M. E. Elbuluk, G. C. Verghese, and J. G. Kassakian, “Sampled-data modeling and digital control of resonant converters,” IEEE Trans. Power Electronics, vol. 3, no. 3, pp. 344–354, Jul. 1988.